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# Time-Optimal Control of an Ultrasonic Sensor

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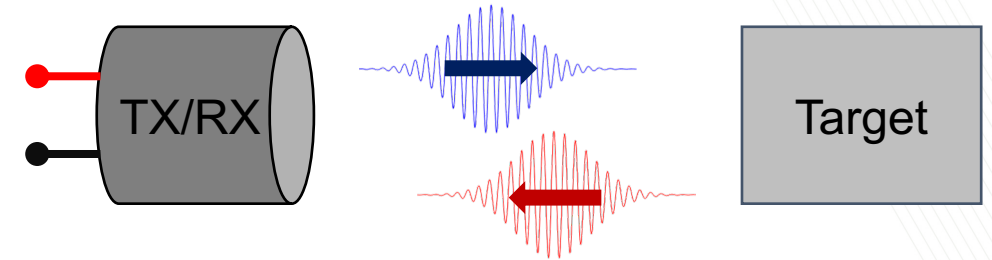
Project Sponsor: Dr. David Magee (Texas Instruments)

# Outline

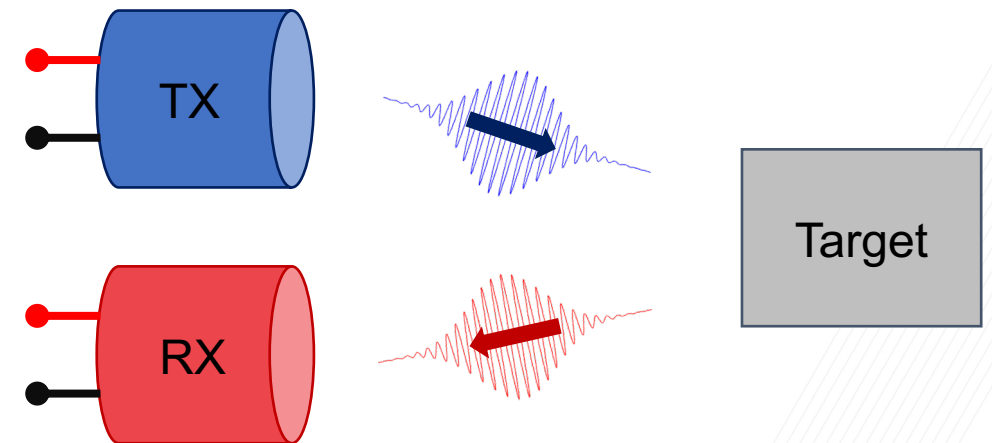
- Operating principles of ultrasonic distance sensors
- Range and resolution limitations of ultrasonic distance sensors
- Improving sensor range and resolution via time-optimal control
- Role of implementation constraints in formulation of the control problem
- Model-free solution to implementation-constrained time-optimal control problem

# Distance Measurement with an Ultrasonic Sensor

- An ultrasonic sensor consists of one or more ultrasonic transducers and a signal processor
- Distance measurement begins with the transmission of a pulse towards a target
- Under suitable conditions, an echo is received after a time delay
  - Target orientation, material composition and transmit pulse strength play major roles
- Transducer voltage waveforms are processed to estimate the time delay and compute distance to the target



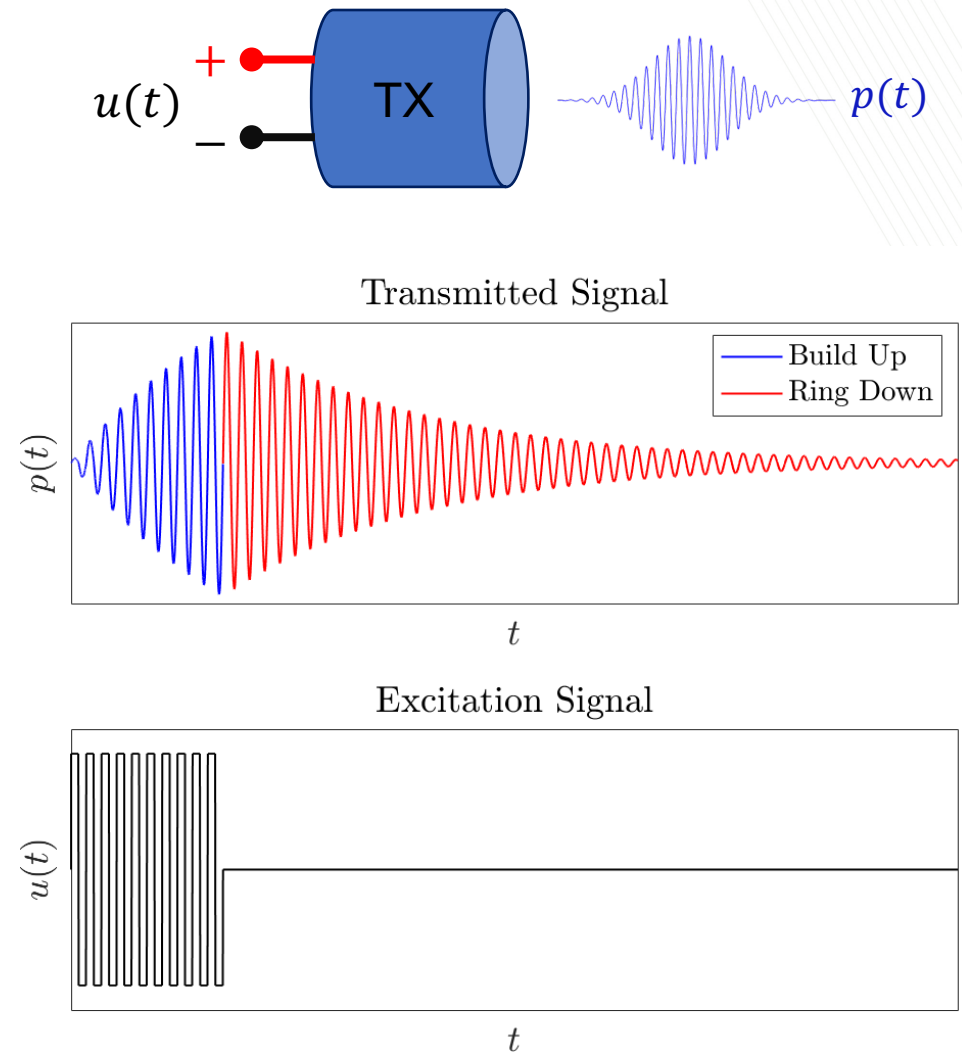
Single-transducer configuration (monostatic)



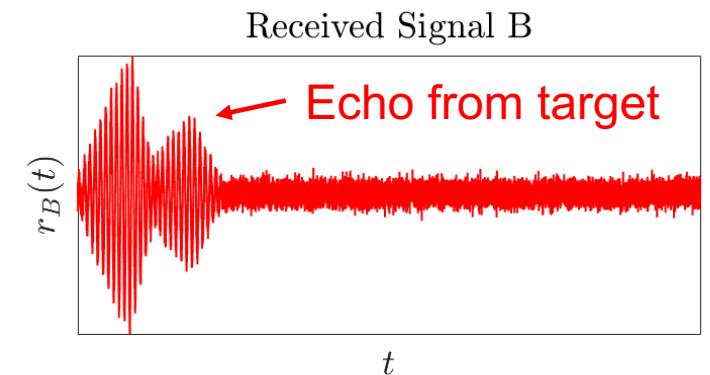
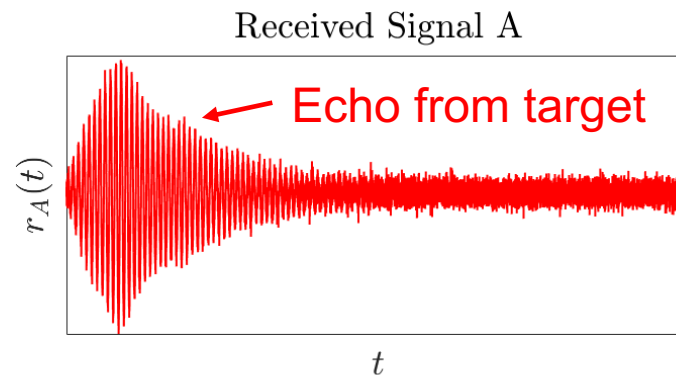
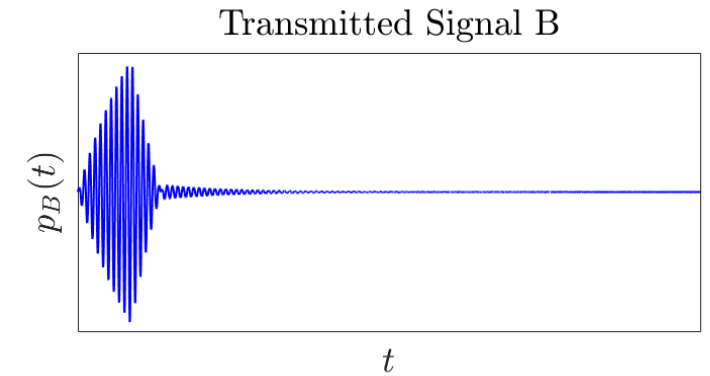
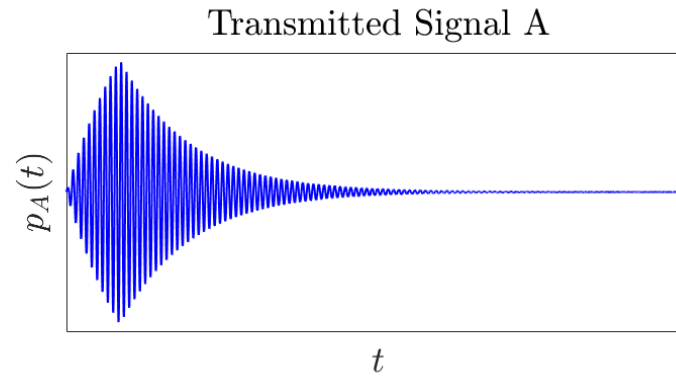
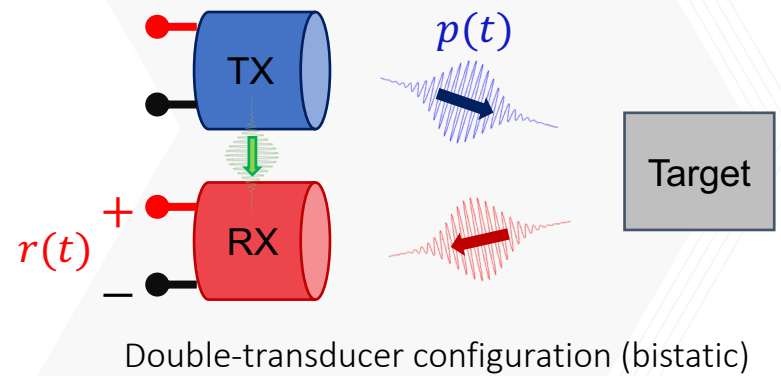
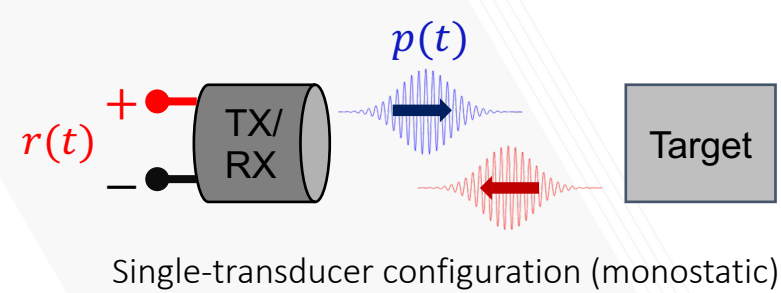
Double-transducer configuration (bistatic)

# Pulse Generation in Piezoelectric Ultrasonic Transducers

- Piezoelectric ultrasonic transducers are poorly damped resonant systems by design
- Typical excitation signal consists of an integer number of pulses at the resonant frequency
  - Low damping ensures high-amplitude pressure wave transmission
    - Necessary to sense distant targets and combat propagation/reflection loss
  - Low damping gives rise to a long ring-down period
    - Ringing has several undesirable consequences
- *Transducer design tradeoff*: Introducing damping decreases ring-down time at the cost of pulse strength



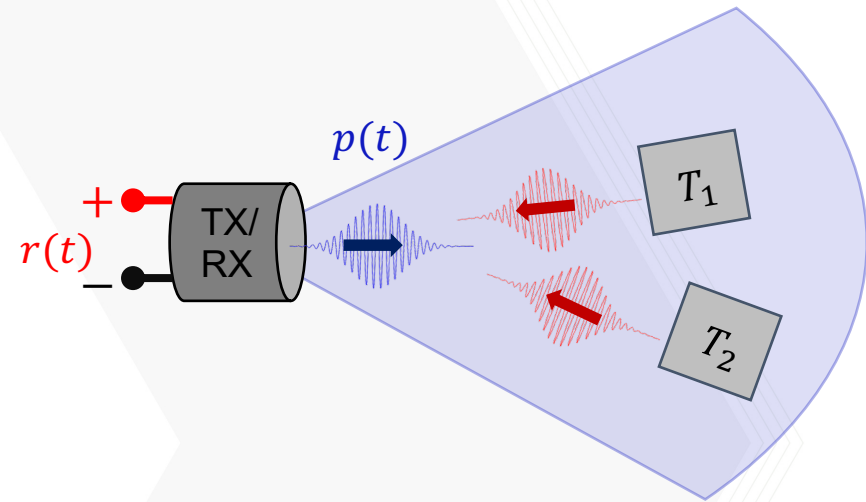
# Issue #1: Ringing Limits Sensing Range



- *Scenario:* Object located close to sensor
- Without ringing suppression, *the received echo is completely obscured by ringing*
  - Clearly, both thresholding and correlation-based detection schemes will fail here
- Note: Transmit activity can appear in  $r_{A/B}(t)$  even in the bistatic case due to *crosstalk (green)*

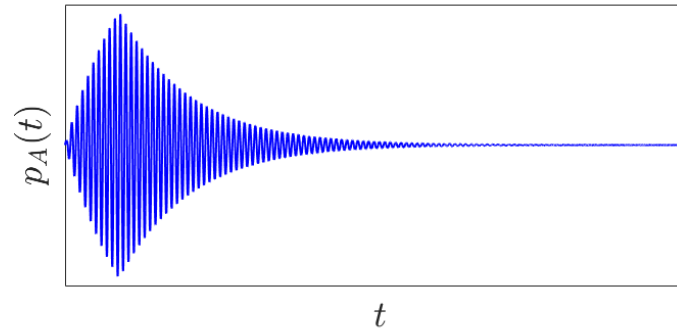


# Issue #2: Ringing Limits Sensing Resolution

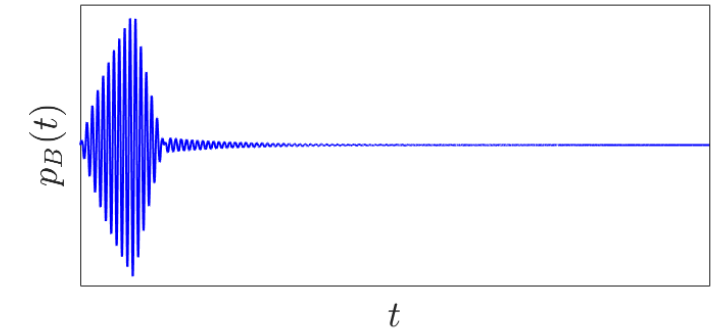


Monostatic configuration shown. Applies equally to bistatic configuration

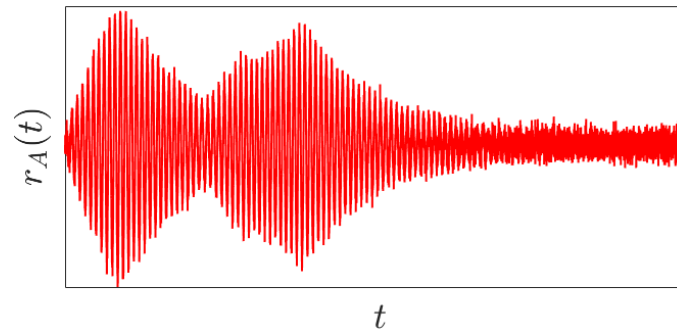
Transmitted Signal A



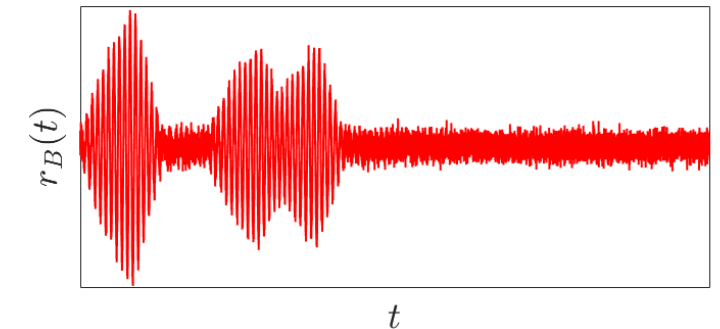
Transmitted Signal B



Received Signal A

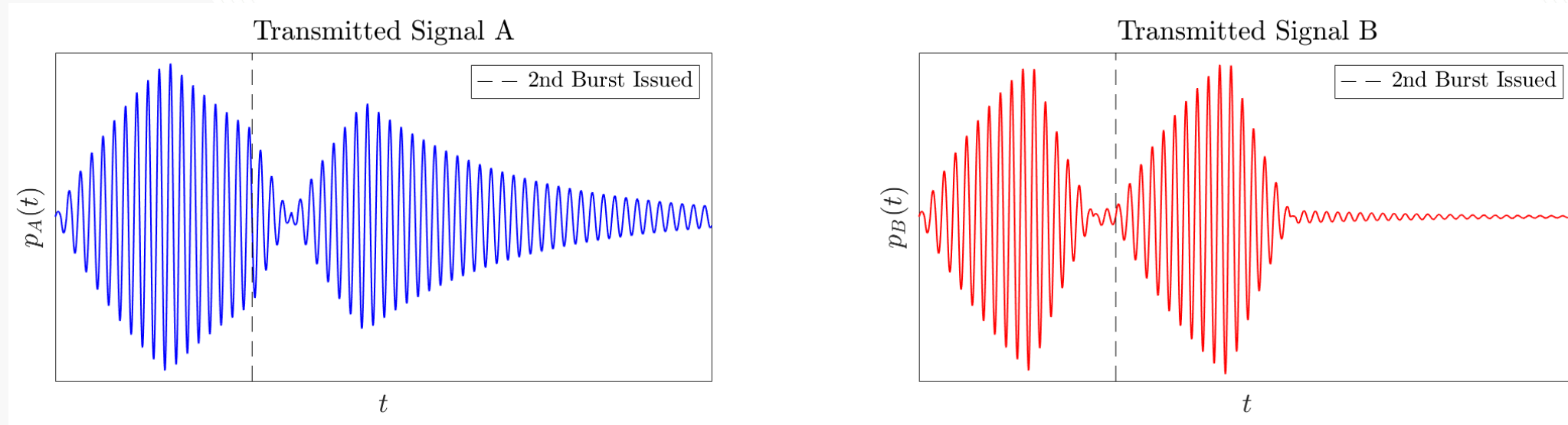


Received Signal B



- *Scenario:* Two distinct targets located close together within field of view
- Transducers have a finite *beam width*. Acoustic energy is transmitted into a conical region of space which may contain multiple objects. Multiple echoes may return
- Without ringing suppression, the *echoes from target  $T_1$  and target  $T_2$  are not separable*

# Issue #3: Ringing Creates Variability in Pulse Transmission



- *Scenario:* Two burst events issued in rapid succession
- This situation could arise if multiple distance measurements are being averaged to reduce the influence of noise
- The amplitude of the second burst pulse can vary significantly depending on the time separation between burst events
  - Can lead to failed object detections when detections are expected

# First Statement of the Control Problem

- Poor damping of piezoelectric ultrasonic transducers is beneficial during pulse transmission, but also gives rise to undesirable ringing
  - We wish to drive the system to rest in minimum time, from its state at the *end of burst*
- The natural solution is time-optimal control. May consider continuous-time first
- Given  $x_0$ , choose  $u(t) \in \mathcal{U}_c$  for  $0 \leq t \leq T$  to
  - minimize  $T$
  - subject to  $\dot{x}(t) = Ax(t) + Bu(t)$
  - $x(0) = x_0, \quad x(T) = 0$
- Possible input constraint set:  $\mathcal{U}_c = \{u : |u| \leq U_{max}\}$
- This problem is treated in textbooks, solution is bang-bang with edges in  $[0, T]$  [1]
- Requires analog circuitry to implement. We favor microcontroller-based implementation

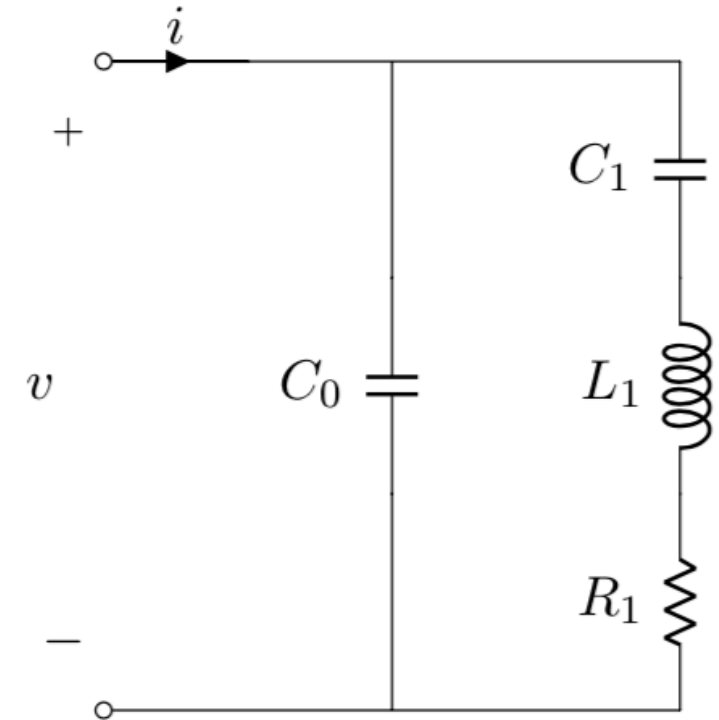


# Reformulating the Control Problem

- If using a microcontroller, placement of edges is restricted to a discrete set of points. Discrete-time time-optimal control is more appropriate:
- Given  $x_0$ , choose  $u[k] \in \mathcal{U}_d$  for  $k = 0, \dots, K - 1$  to
  - minimize  $K$
  - subject to  $x[k + 1] = A_d x[k] + B_d u[k]$
  - $x[0] = x_0, \quad x[K] \in \mathcal{S}$
- Connections between continuous and discrete problems:  $A_d = e^{AT_s}, \quad B_d = \int_0^{T_s} e^{At} B dt, \quad T \sim KT_s$
- $\mathcal{U}_d = \{u : |u| \leq U_{max}\}, \quad \mathcal{S}_0 = \{0\}$ 
  - Solution via convex optimization in [2]; does not account for actuator quantization
- $\mathcal{U}_d = \{\pm(q/Q)U_{max}, q = 0, 1, \dots, Q\}, \quad \mathcal{S}_\epsilon = \{x : \|x\| \leq \epsilon\}$ 
  - Accounts for discrete-time and discrete signal amplitude constraints [3]
  - $\mathcal{U}_d$  is a finite set, therefore searchable. Special case:  $Q = 1 \Rightarrow \mathcal{U}_d = \{-U_{max}, 0, U_{max}\}$
  - Search-based solution allows us to shed dependence on a model

# Inspiration: Time-Optimal Control of the Harmonic Oscillator

- Could exhaustively search over all possible control sequences such that  $u[k] \in \{-U_{max}, 0, U_{max}\}$ ,  $k = 0, 1, \dots, K - 1$
- *Can we use intuition to find a more intelligent (faster) solution?*
- Transducer response resembles that of a second-order damped harmonic oscillator
  - Due to *poor damping*, studying the *undamped* oscillator is informative
- The continuous-time time-optimal control to drive the state of the harmonic oscillator from an initial condition to the origin is bang-bang at the resonant frequency
  - Easy to realize (at least approximately) using a microcontroller
  - Reduces size of search space!



Butterworth Van-Dyke model of an ultrasonic transducer –  $RLC$  branch represents a mechanical oscillator

# Inspiration: Time-Optimal Control of the Harmonic Oscillator

- The undamped harmonic oscillator is described by:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & \omega \\ -\omega & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \quad , |u(t)| \leq 1$$

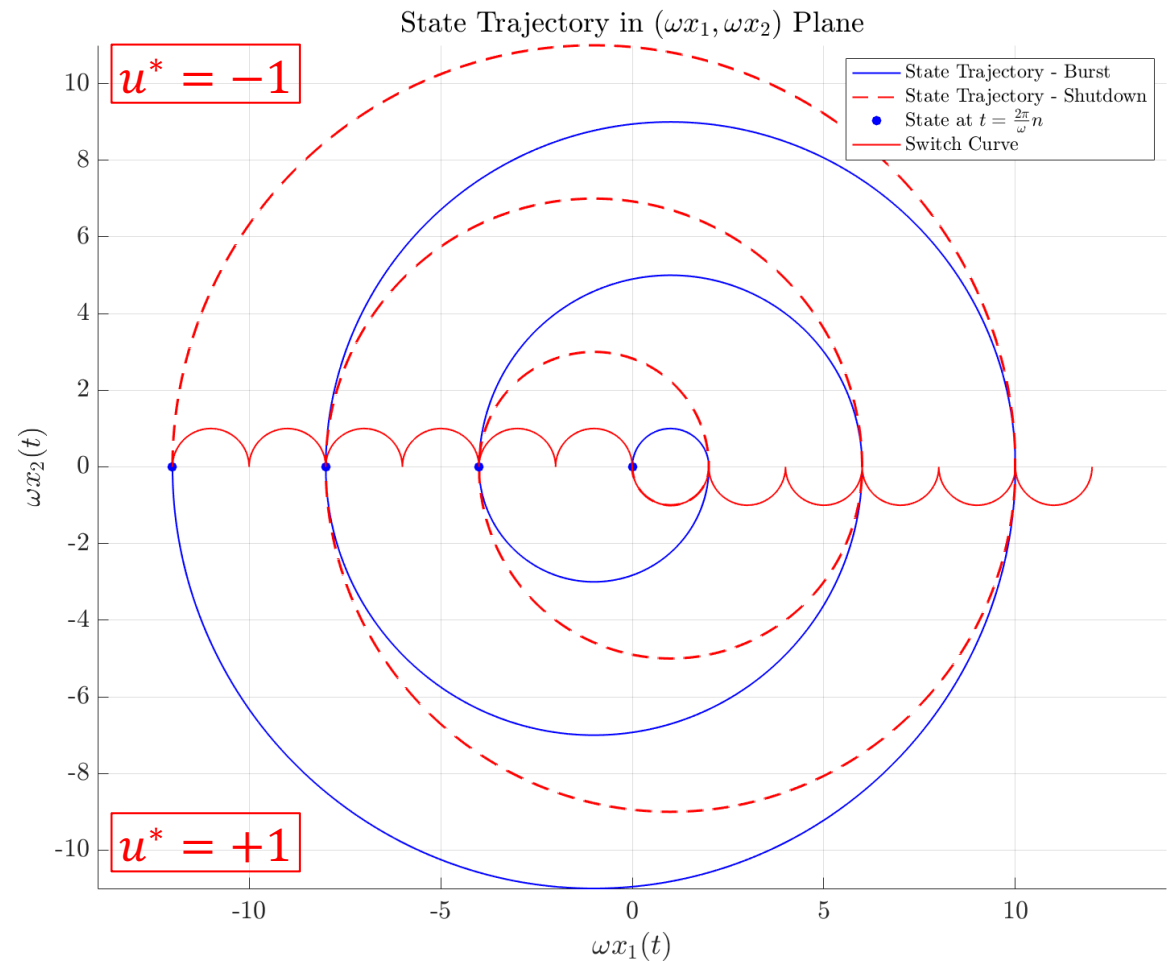
- It can be shown that when subjected to the input:

$$u(t) = \begin{cases} \text{sign}(\sin(\omega t)) & , \text{for } N_E \text{ cycles} \\ 0 & , \text{otherwise} \end{cases}$$

- The (scaled) state after the  $n^{\text{th}}$  of  $N_E$  cycles is given by:

$$\omega x \left( \frac{2\pi}{\omega} n \right) = \begin{bmatrix} -4n \\ 0 \end{bmatrix} \quad , n = 0, \dots, N_E$$

- The time origin for control is the end of burst; initial conditions for control are located on the switch curve
- The first control action must therefore be -1 for  $\pi/\omega$  seconds, irrespective of the resonant frequency!



# Algorithm: Model-Free Shutdown

- We propose finding the time-optimal control, subject to implementation constraints, via a brief, one-time power-on calibration procedure:

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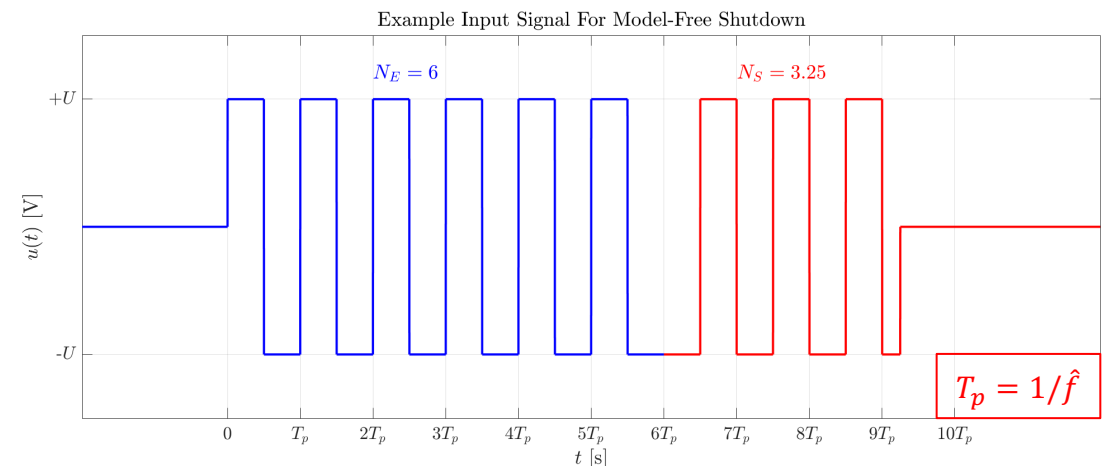
 $\hat{f}$  = Estimated resonant frequency in Hz
 $N_{s,prev} = 0$ 
for  $N_E = 1 : 10$ 
     $t_{min} = +\infty$ 
    for  $N_S = N_{s,prev} : 1/16 : N_E$ 
        Issue  $N_E$  burst pulses at  $\hat{f}$  Hz
        Issue  $N_S$  shutdown pulses at  $\hat{f}$  Hz
         $t_s$  = Measured settling time
        if  $t_s < t_{min}$ 
             $t_{min} = t_s$ 
             $N_{s,prev} = N_S$ 
        end if
    end for
    Record  $N_{s,prev}$  in Table
end for
    
```

- In the worst case (if  $N_S = 0$  always recommended), 890 burst-and-measure trials are conducted
- If the sample table on the right is produced, 370 burst-and-measure trials are conducted

Sample Table Constructed by Algorithm

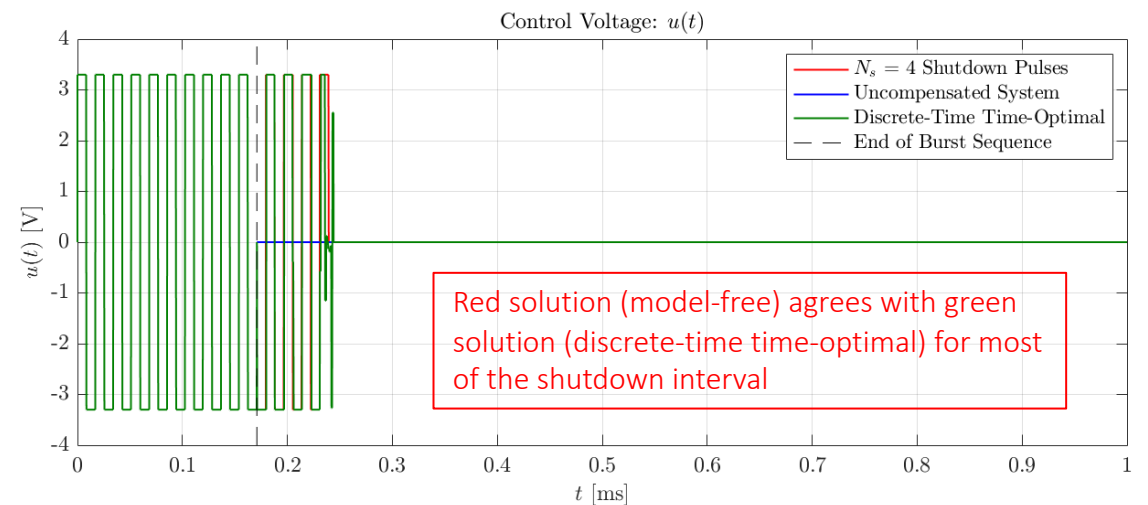
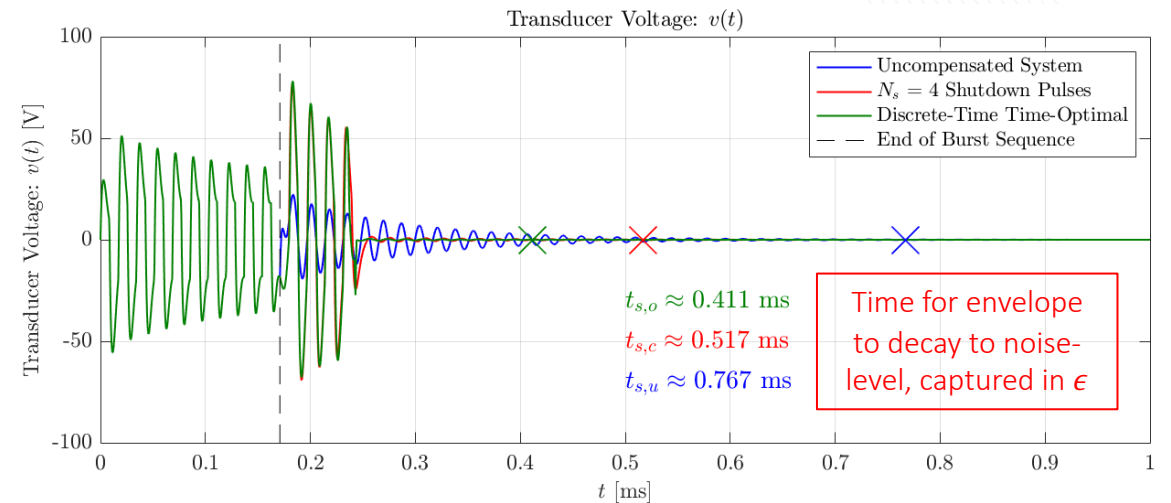
$N_E$	$N_S$	$N_E$	$N_S$
1	2	6	4
2	2.5	7	4.5
3	2.5	8	5
4	3	9	5.5
5	3.5	10	5.5

Sample Candidate Excitation Sequence



# Results and Comparison: Model-Free Shutdown

- Simulation model developed for a 58 kHz transducer + impedance matching network
- **Blue:** No shutdown action
- **Green:** Discrete-time time-optimal shutdown
- **Red:** Model-free shutdown
  - $\mathcal{U}_d = \{-U_{max}, 0, U_{max}\}$ ,  $\mathcal{S}_\epsilon = \{x : \|x\| \leq \epsilon\}$
- Discrete-time time-optimal solution provides a *benchmark*
  - A precise model and a departure from current hardware are required for implementation
- Table construction is done once at power-up. At runtime, only a table *lookup* is required. No expensive computation!





# Summary, Conclusion

- To enhance sensor range and resolution, short pulses are preferred over long pulses
- Time-optimal control is employed to produce the *shortest possible* pulses by suppressing ringing after burst
- Practical implementation constraints influence the formulation of the control problem
- A simple, model-free solution to the properly reformulated control problem is presented
  - This solution is *time-optimal*, subject to practical implementation constraints
  - This solution is model-free. Not susceptible to model parameter error
  - This solution is inspired by 2<sup>nd</sup> order theory, but was shown to perform well even on a 4<sup>th</sup> order system

# References

- [1] M. Athans and P. Falb, *Optimal Control: An Introduction to the Theory and Its Applications*. McGraw-Hill, 1966.
- [2] L. Bako, D. Chen, and S. Lecoeuche, “A numerical solution to the minimum-time control problem for linear discrete-time systems,” *CoRR*, 2011
- [3] J. V. Outrata, “On the minimum time problem in linear discrete systems with the discrete set of admissible controls,” *Kybernetika*, vol. 11, no. 5, pp. 368–374, 1975.

**Thank you for this opportunity!**

# Backup Slide: Model Details, Energy Removal Perspective

